# Outline

- Wiener Filters: Introduction
- FIR Wiener Filter
- Orthogonality Principle
- Causal IIR Wiener Filter
- Non-causal IIR Wiener Filter

### Wiener Filters: Introduction

- Now we are considering the problem of estimating a signal in the presence of noise
- In a lot of cases additive noise model is used

x[n] = s[n] + w[n],

where x[n] is the measured signal, s[n] is the desired signal, and w[n] is the conteminating noise

• Let us filter the measurement x[n]

$$y[n] = \mathcal{H}\{x[n]\}$$

# Wiener Filters: Introduction (Cont.)

- Depending on the application we want different y[n]'s
- In general we have

$$y[n] = s[n+D]$$

- D = 0 corresponds to usual filtering, the only goal is to get rid of the noise
- D > 0 corresponds to prediction, we are trying to reach a guess of future values

## Wiener Filters: Introduction (Cont.)

- We use minimum mean-squared error as the optimality criterion
- Assume that all random signals involved are zero mean and wide sense stationary



Figure 12.7.1 Model for linear estimation problem.

### **FIR Wiener Filter**

- We restrict the optimum linear filter to be FIR
- Then the filtered signal is

$$y[n] = \sum_{k=0}^{M-1} h[k]x[n-k]$$

• Our criterion: minimum mean-squared error (MMSE)

$$h_{opt} = \arg\min_{h} \mathbb{E}\{|d[n] - \sum_{k=0}^{M-1} h[k]x[n-k]|^2\}$$

• Taking the derivative with respect to h and equating to zero we obtain

$$\sum_{k=0}^{M-1} h[k]R_x[l-k] = R_{dx}[l] \qquad l = 0, 1, \dots, M-1$$

• A generalized version of normal equations

• We can write these equations in matrix form

$$\Gamma_M \boldsymbol{h} = \boldsymbol{r}_d$$

resulting in the solution

$$oldsymbol{h}_{opt} = \Gamma_M^{-1}oldsymbol{r}_d$$

- The resulting MMSE can be found by substituting this optimum filter in the error expression
- Consider the expression for the MMSE

$$\epsilon_{M} = E\{|d[n] - \sum_{k=0}^{M-1} h[k]x[n-k]|^{2}\} \\ = E\{|d[n]|^{2} + \sum_{k=0}^{M-1} \sum_{k'=0}^{M-1} h[k]h[k']^{*}x[n-k]x^{*}[n-k'] \\ - \sum_{k=0}^{M-1} d[n]h^{*}[k]x^{*}[n-k] \\ - \sum_{k=0}^{M-1} d^{*}[n]h[k]x[n-k]\}$$

• But we have the following equality for optimum  $\boldsymbol{h}$ 

$$\sum_{k=0}^{M-1} h[k]R_x[l-k] = R_{dx}[l]$$

• Substituting this expression we obtain

MMSE<sub>M</sub> = 
$$\sigma_d^2 - \sum_{k=0}^{M-1} h_{opt}[k] R_{dx}^*[k]$$

• Substituting  $\boldsymbol{h}_{opt} = \Gamma_M^{-1} \boldsymbol{r}_d$  we obtain

$$\mathrm{MMSE}_M = \sigma_d^2 - \boldsymbol{r}_d^{\mathrm{H}} \Gamma_M^{-1} \boldsymbol{r}_d$$

- Filtering noise: we have x[n] = s[n] + w[n] and d[n] = s[n]
- We can also assume that the signal s[n] and noise w[n] are independent

$$R_x[k] = R_s[k] + R_w[k]$$

$$R_{dx}[k] = R_x[k]$$

• Substituting these normal equations become

$$\sum_{k=0}^{M-1} h[k] \{ R_s[l-k] + R_w[l-k] \} = R_s[l]$$

• Prediction: we have

$$d[n] = s[n+D]$$
$$R_{dx}[k] = R_s[l+D]$$

• The normal equations become

$$\sum_{k=0}^{M-1} h[k] \{ R_s[l-k] + R_w[l-k] \} = R_s[l+D]$$

## **Orthogonality Principle**

• Remember the normal equations

$$\sum_{k=0}^{M-1} h[k]R_x[l-k] = R_{dx}[l] \qquad l = 0, 1, \dots, M-1$$

• Rearranging terms

$$R_{dx}[l] - \sum_{k=0}^{M-1} h[k]R_x[l-k] = 0$$

• This can be written as

$$\mathbf{E}\{e[n]x^*[n-l]\} = 0$$

where

$$e[n] = d[n] - \sum_{k=0}^{M-1} h[k]x[n-k]$$
$$e[n] = d[n] - \hat{d}[n]$$

### **Orthogonality Princple**

• Orthogonality principle is best explained graphically



![](_page_10_Figure_3.jpeg)

#### **Causal IIR Wiener Filter**

• We now allow the filter to have infinite length

$$y[n] = \sum_{k=0}^{\infty} h[k]x[n-k]$$

• All steps are similar except the summation runs to  $\infty$  leading to normal equations

$$\sum_{k=0}^{\infty} h[k]R_x[l-k] = R_{dx}[l] \qquad l \ge 0$$

• The MMSE for the optimum filter coefficients is

$$\mathrm{MMSE}_M = \sigma_d^2 - \boldsymbol{r}_d^{\mathrm{H}} \Gamma_M^{-1} \boldsymbol{r}_d$$

- We cannot use matrix inversion to solve these normal equations since they are of infinite length
- We cannot directly use z-transform neither because the equations are for l>0
- Remember we had this half-infinite summation problem in innovations representation, we will use a similar trick
- The optimum filter h is applied to x, but x can be obtained from white noise with innovations representation, and the inverse of this produces white noise from x

• Therefore we can decompose *h* into two parts: whitening filter and optimum filter

$$x[n] \to H \to y[n]$$
$$x[n] \to \text{Whitening} \to i[n] \to Q \to y[n]$$

• We now have

$$y[n] = \sum_{k=0}^{\infty} q[k]i[n-k]$$

producing

$$\sum_{k=0}^{\infty} q[k]R_i[l-k] = R_{di}[l] \qquad l \le 0$$

• Using the fact that *i* is white noise

$$q[l] = \frac{R_{di}[l]}{R_i[0]} = \frac{R_{di}[l]}{\sigma_i^2}$$

- To calculate the z-transform of h let us calculate first the z-transform of q and then the z-transform of whitening filter. Product of two will give the z-transform of h
- The z-transform of q

$$Q(z) = \sum_{k=0}^{\infty} q[k] z^{-k}$$
$$= \frac{1}{\sigma_i^2} \sum_{k=0}^{\infty} R_{di}[k] z^{-k}$$

• This is the right sided z-transform of  $R_{di}$ 

$$[\Gamma_{di}(z)]_{+} = \sum_{k=0}^{\infty} R_{di}[k] z^{-k}$$

• We can calculate this as follows: start with the whitening filter

$$i[n] = \sum_{k=0}^{\infty} v[k]x[n-k]$$

• We have

$$R_{di}[k] = E\{d[n]i^*[n-k]\}$$
  
=  $\sum_{m=0}^{\infty} v[m]E\{d[n]x^*[n-m-k]\}$   
=  $\sum_{m=0}^{\infty} v[m]R_{dx}[k+m]$ 

• Substituting this into definition of z-transform

$$\Gamma_{di}(z) = \sum_{k=-\infty}^{\infty} \{\sum_{m=0}^{\infty} v[m] R_{dx}[k+m] \} z^{-k}$$

• Interchanging order of summations

$$\Gamma_{di}(z) = \sum_{m=0}^{\infty} v[m] \sum_{k=-\infty}^{\infty} R_{dx}[k+m] z^{-k}$$

• Change of variable k' = k + m

$$\Gamma_{di}(z) = \sum_{m=0}^{\infty} v[m] z^m \sum_{k'=-\infty}^{\infty} R_{dx}[k'] z^{-k'}$$
$$= V(z^{-1}) \Gamma_{dx}(z)$$

• Finally combining with whitening filter

$$H_{opt}(z) = Q(z)V(z)$$

where

$$Q(z) = \frac{1}{\sigma_i^2} \left[ V(z^{-1}) \Gamma_{dx}(z) \right]_+$$

#### Non-causal IIR Wiener Filter

• We now allow the filter to multiply future values as well

$$y[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$$

• All steps are similar except the summation runs to  $\infty$  leading to normal equations

$$\sum_{k=-\infty}^{\infty} h[k]R_x[l-k] = R_{dx}[l] \qquad l\epsilon Z$$

• We can use z-transform directly in this case

$$H(z) = \frac{\Gamma_{dx}(z)}{\Gamma_{xx}(z)}$$